Open-Circuit Fault Detection Technique for Fault-Tolerant Bridgeless Boost Rectifier

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Abstract

Unexpected failure in one of the switches in a converter will lessen the performance of a system and may even prompt to a system failure. Therefore, to reduce the failure rate and to prevent unscheduled shutdown, a real-time fault detection is necessary. In this paper, a new technique to identify the exact location of the faulty switch based on single-phase bridgeless boost converter is proposed. The faulty switch is detected by analyzing the boost inductor current, switching mode and cycle of the input power supply. The information of the faulty switch will be used to redirect the current to any available path. Therefore, an uninterruptable and reliable power supply could be developed. The analysis of the proposed fault detection scheme is done by using PSIM simulation software.

Keywords: Bridgeless Boost Rectifier, Fault Detection, Fault-Tolerant, PowerSim, Faulty Switches.

1. Introduction

Recently, power converter is amongst the devices which is in high demands in market trend. This is due to high quantities of hybrid and electric cars release in market which require redundant features in this era of technology. However, power semiconductor failure known to be one of the main factor to the converter destruction. Consequently, evolution of fault-tolerant systems has been presented in previous research [1-3]. Fault-tolerant control methods and open switch fault detection are popular in those interested topics in research.

With a specific end goal to change converter configuration and evade from subordinate harm in a fault-converter, a fault detection algorithm ought to be reasonable which is fast recognition, basic configuration and exact estimation [4]. Basically, the major occurrence of fault can be divided into two types, which are open-circuit fault (OCF) and short-circuit fault (SCF) [5]. SCF can occurs due to various conditions such as cable leakage, switching devices failure and auxiliary power supply breakdown. SCF may contributes to power system breakdown. Along these lines, equipment based protection, for example, fuses and circuit breakers can be used as protection to keep the fault occurrence from ending up more harmful. Nevertheless, OCF may not resorting to a damage of a system but rather it might cause an unusual process or execution corruption. The OCF might occurs due to lift of the bonding wire, driver failure or short-circuit fault-induced rupture [5].

Few researchers mentioned the importance of redundancy approached in power converters due to converter failure that leads to the troublesome often requiring additional repairs, and crucial implementation of destructive converters can be harmful [6]. The redundancy approached mentioned as well-known fault-tolerant features for the power converters. Fault-tolerant features function is to ensure the continuity of power of a converter. Subsequently, the fault-tolerant studies development introduces analysis of fault detection and protection methods (FDPM) that requires literature studies on the existing methods and techniques.

There have been many FDPM proposed over recent years which have their own advantages and drawbacks. The most popular FDPM are applied to modular multilevel converter that have high number of switches. The FDPM method introduced to this converter topologies are capacitor voltage estimation which presents a new voltage balancing control technique with fault detection and fault tolerant capability based on capacitor voltage estimation instead of direct measurement. This method has high complexity of the circuit with fault tolerant capability added [7]. In [8], measured current behavior is compared to the expected current behavior to detect the fault occurrence and its location due to gate signals and the sign of arm current. This method is simple and fast detection of fault. Besides, model predictive control method checking the voltage errors between the measured arm voltages and the estimated ones in the former control cycle [9]. This method can only detect and isolate the single open-circuit fault and it fails when multiple open-circuit faults occur simultaneously. Based on analysis on these literature studies, the detection of FDPM approached on modular multilevel converters are average. Except for one method which is based on the deviation's magnitude and current flow direction. This method has fast speed detection magnitude which is less than 20ms.

The proposed topology in this research is single-phase bridgeless boost rectifier. The focus on fault detection and protection methods in this study are narrowed down to fault detection techniques that must be simple enough to be implemented, accurate, and fast. There are some methods that have been introduced in recent papers which applies on single-phase converters and have fast fault detection speed which are source-to-ground voltage measurement of the floating MOSFET, drain to source voltage monitoring and diode-voltage signature [10-11]. However, there is no fault detection technique that focused on bridgeless converter or more specifically bridgeless boost rectifier. In this way, with the inspiration of having simple circuit configuration, ability to reroute the current

path to other healthy switches after fault occurrence at any instant, a new technique to detect the exact location of the faulty switch for fault-tolerant bridgeless boost rectifier (FTBBR) is proposed.

2. Fault Tolerant Bridgeless Boost Rectifier

In this section, the FTBBR circuit configuration is explained. FTBBR is able to continuously supplying power to the load even if two out of four switches are open-circuited. The FTBBR circuit is shown in Figure 1 and the detailed circuit operation can be found in [12]. When OCF is occurred, the current flow will be channeled to any possible path by turning on any possible switch. The channeling of the current is called as current reroute path (CRP).
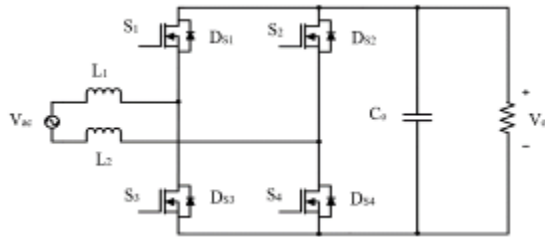

Fig. 1: The FTBBR circuit configuration [12]

The occurrence of fault can be characterized into four states as follows:

1. First state is when the upper and lower horizontal parallel pairs (S_1 and S_2 or S_3 and S_4) or right and left vertical parallel pairs (S_1 and S_3 or S_2 and S_4) are open-circuited. In this state, the FTBBR will have one CRP.
2. Second state is when the OCF take placed at one of the switches. In particular, the FTBBR has two alternative paths for CRP.
3. Third state is when the OCF take placed at the diagonal pairs (S_1 and S_4 or S_2 and S_3). The FTBBR will be malfunctioned as there is no route for the current to flow. However, the converter will be operated as a full-wave uncontrolled rectifier i.e. by using the body diodes of the MOSFETs.
4. The fourth state is when only one switch is functioning. This state is forbidden because two switches are required to be turned-on in order to enable the current to flow.

The FTBBR is able to continuously operate as a full-wave uncontrolled rectifier over the diodes if the OCF occurs under the third and fourth state or even all switches are open circuited. Table 1 shows the summary of the fault possibility, T_X and the state of the converter operation, where X is the possibility condition from A to P.

Table 1: Fault Possibility Table

Fault possibility	S_1	S_2	S_3	S_4	Converter operation
T_A	0	0	0	0	Full-wave uncontrolled rectifier
T_B	0	0	0	1	Full-wave uncontrolled rectifier
T_C	0	0	1	0	Full-wave uncontrolled rectifier
T_D	0	0	1	1	CRP 1
T_E	0	1	0	0	Full-wave uncontrolled rectifier
T_F	0	1	0	1	CRP 3
T_G	0	1	1	0	Full-wave uncontrolled rectifier
T_H	0	1	1	1	CRP 1 and CRP 3
T_I	1	0	0	0	Full-wave uncontrolled rectifier
T_J	1	0	0	1	Full-wave uncontrolled rectifier
T_K	1	0	1	0	CRP 4
T_L	1	0	1	1	CRP 1 and CRP 4
T_M	1	1	0	0	CRP 2
T_N	1	1	0	1	CRP 2 and CRP 3
T_O	1	1	1	0	CRP 2 and CRP 4
T_P	1	1	1	1	All paths

2.1. Generalized Steady State FTBBR Operation

There are four current reroute paths (CRP) for FTBBR. Each CRP has four modes of operation which is same for all other paths. To additionally simplify the analysis, it is presumed that all components and devices are ideal.

1. CRP 1, OCF occurs at upper horizontal pair (S_1 And S_2)
   At positive cycle, the inductor charges L_1 and L_2 and current flows through S_3 and D_4 as shown in Figure 1 (positive charging mode). Meanwhile, the output capacitor, C_O is discharged through the load. The power from the input is transferred to the load through D_1 and D_4 when S_3 is turned off (positive discharging mode).
   At negative cycle, the current charges inductor L_1 and L_2 and flows through D_3 and when S_4 is turned-on (negative charging mode). This mode is same as defined in positive charging mode. The input power is transferred to the load through D_2 and D_3 when S_4 is turned-off (negative discharging mode). This mode is same to the positive discharging mode.
   The current flows through similar path for all CRPs during positive and negative cycles at discharging mode,. Therefore, only the charging stage will be clarified from this point onward.
2. CRP 2, OCF occurs at lower horizontal pair (S_3 and S_4)
   At positive cycle, both L_1 and L_2 will be charged through D_1 when S_2 is turned-on. The output capacitor, C_O is discharged through the load. At negative half cycle, L_1 and L_2 are charged when S_1 is turned-on where the current flows through D_{S2}.
3. CRP 3, OCF occurs at left vertical pair (S_1 and S_3)
   L_1 and L_2 are charged when current flows through D_{S1} and S_2 during positive cycle charging stage while L_1 and L_2 are charged when current flows through S_4 and D_{S3} during negative cycle charging stage.
4. CRP 4, OCF occurs at right vertical pair (S_2 and S_4)
   The procedure of the occupied switches is same with CRP 3 but different for the exchange of the input power supply leg. At positive cycle charging stage, the current is flows through S_3 and D_{S4} and charged L_1 and L_2 while during negative cycle charging stage the current will flow through D_{S2} and S_1 and charged L_1 and L_2.
   The switches configuration will be in the form of several combinations of more than one CRP if only one switch is open circuited.

3. Open Circuit Fault Detection Technique

In order to identify the location of the faulty switch is by observing output voltage, cycle of the input supply and inductor current. The output voltage is set to be at 400 V. First the level of the output voltage will be check. If it is less than 400 V, it shows that there is a problem with one of the switch. Then the cycle of the input supply need to be determined either it is during positive or negative cycle. After that the value of I_L need to be measured to ensure that the inductor current is zero. Once the inductor current is measured, the location of the faulty switch could be determined. Figure 2(a-d) show the flowchart to detect the faulty switch for all CRPs.
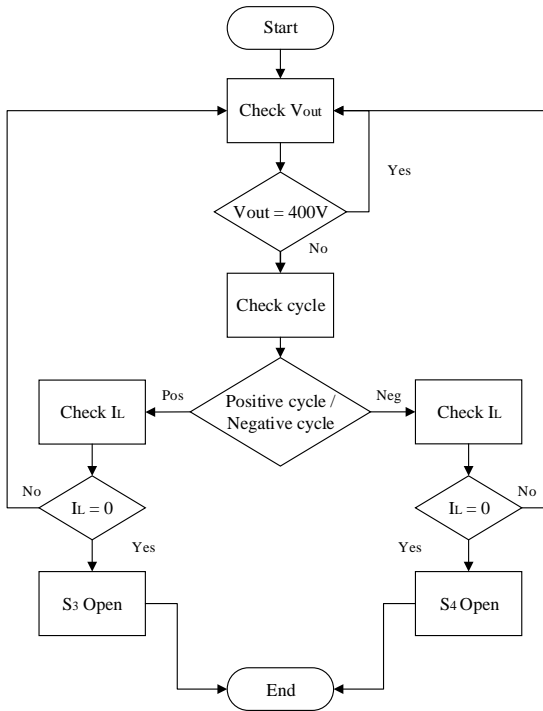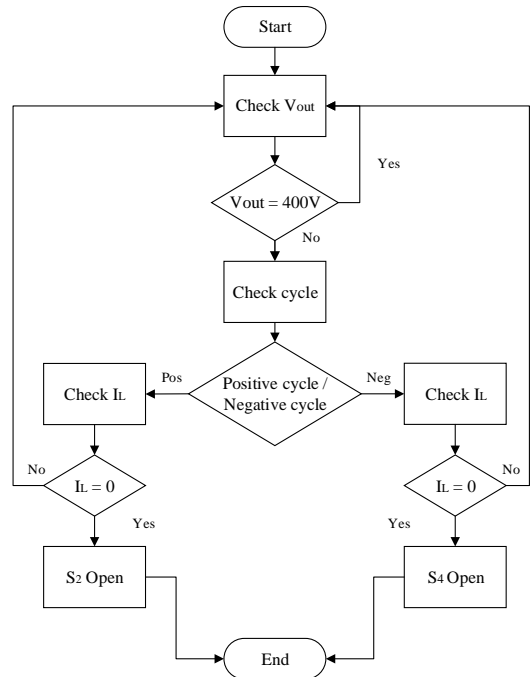
Fig. 2(a): Flowchart of CRP 1
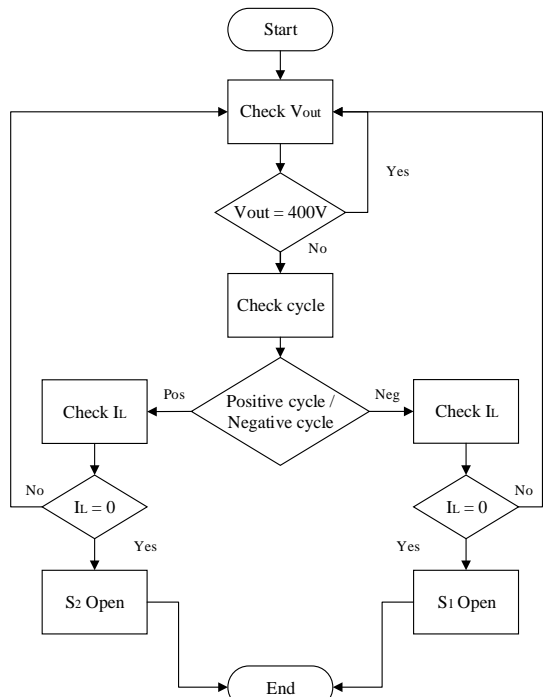

Fig. 2(c): Flowchart Of CRP 3
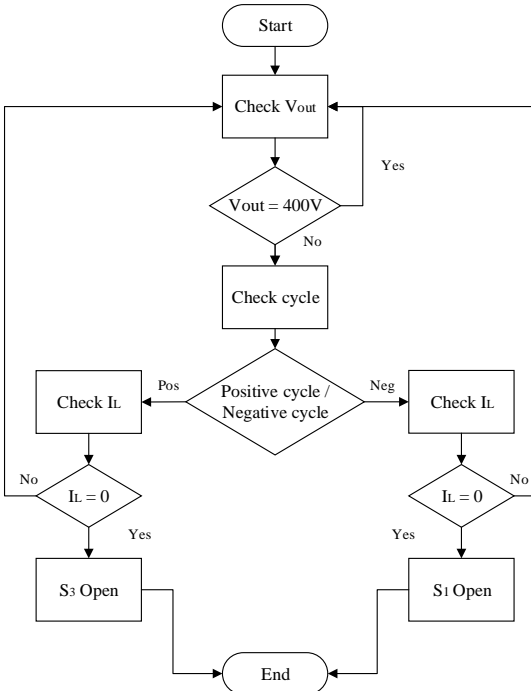

Fig. 2(b): Flowchart of CRP 2

Fig. 2(d): Flowchart Of CRP 4

4. Results and Discussion

To verify the effectiveness of the proposed detection technique, the FTBBR circuit is simulated by using PSIM simulation software. The waveforms of the output voltage and inductor current after the fault occurrence are the main reference to detect the location of the faulty switch. The fault is simulated by turning off the gate signal to the switch which needed to be simulated as OCF. Figure 3 shows an example of the output voltage and the inductor current waveforms during fault occurrence. The output voltage will drop and the inductor current will become zero at either positive or negative cycle.
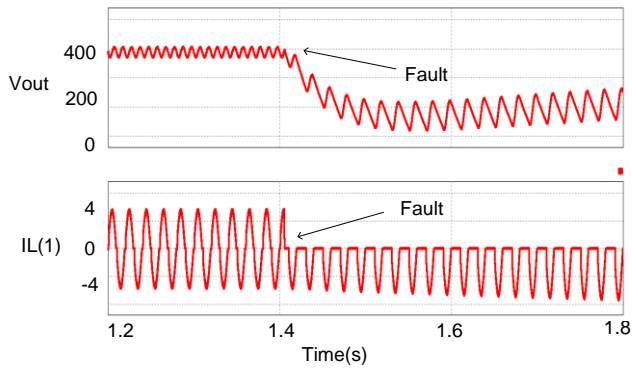
Fig. 3: The waveforms of the output voltage and inductor current before and after fault.

To further strengthen the analysis, the gate signals of the switches are also included to show the converter is running in which CRP before the fault is occurred. Figure 4(a-h) show the waveforms of output voltage, inductor current and gate voltage for all operation modes at positive cycle and negative cycle. Each cycle represents inductor current flow into a switch in particular mode of operation.
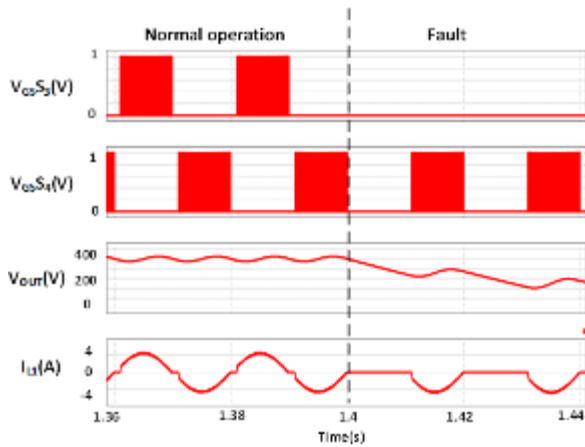

Fig. 4(a): CRP 1 when fault occurred at positive cycle
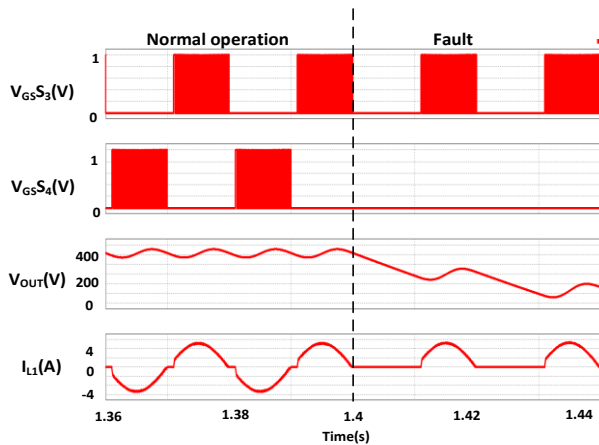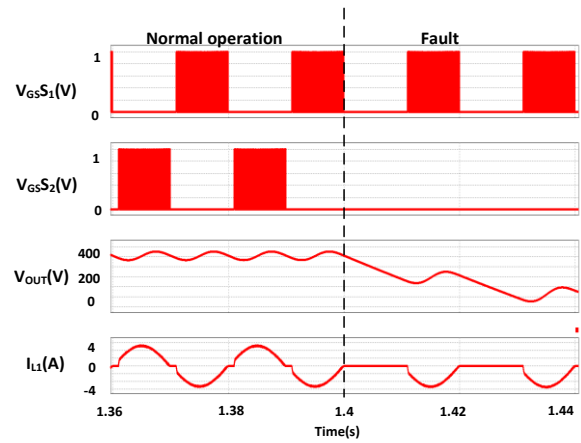

Fig. 4(b): CRP 1 when fault occurred at negative cycle

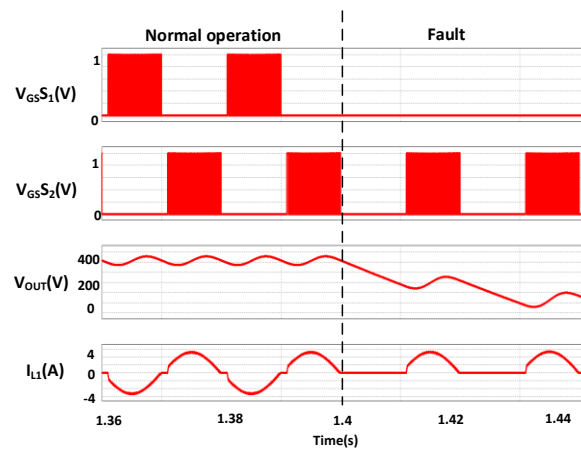**Fig. 4(c):** CRP 2 when fault occurred at positive cycle

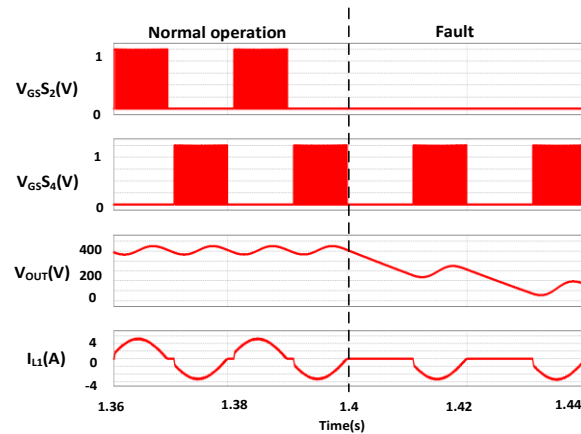**Fig. 4(d):** CRP 2 when fault occurred at negative cycle
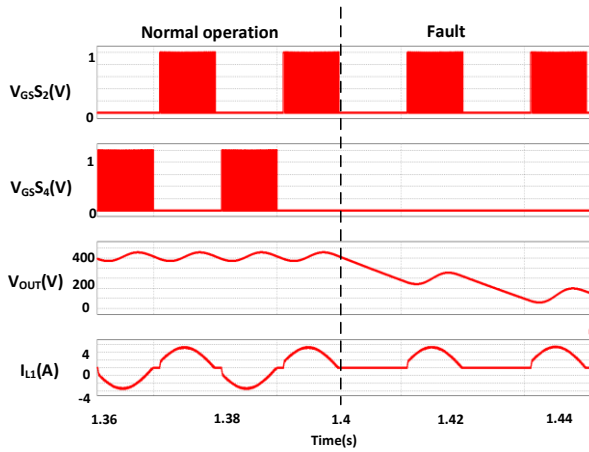

Fig. 4(e): CRP 3 when fault occurred at positive cycle
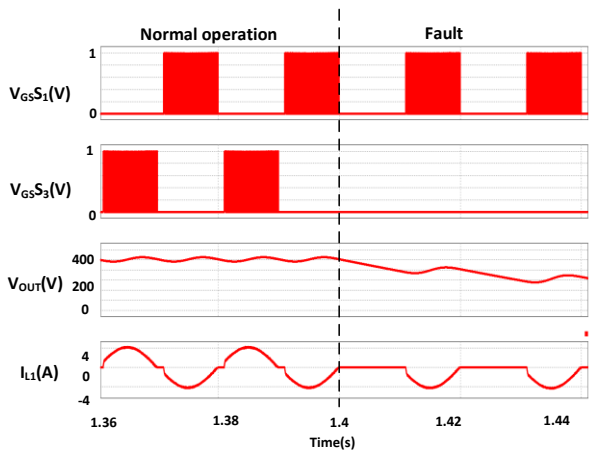
Fig. 4(f): CRP 3 when fault occurred at negative cycle

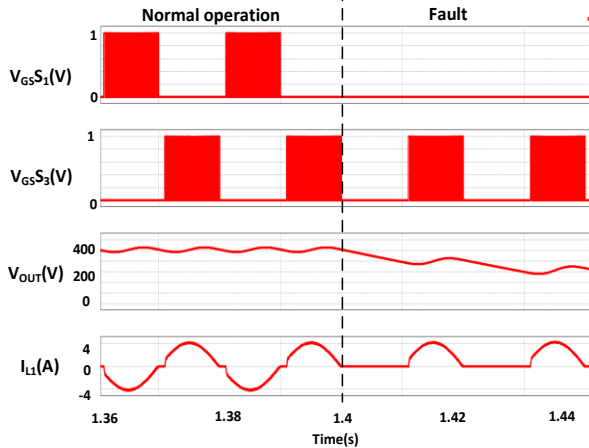**Fig. 4(g):** CRP 4 when fault occurred at positive cycle

Fig. 4(h): CRP 4 when fault occurred at negative cycle

Table 2(a): CRP 1 (Fault at Positive Cycle)

	NORMAL			FAULT		
POSITIVE CYCLE						
MODE	**IL1**	**VGS3**	**VGS4**	**IL1**	**VGS3**	**VGS4**
Charge	+	1	0	0	0	0
Discharge	+	0	0			
NEGATIVE CYCLE						
MODE	**IL1**	**VGS3**	**VGS4**	**IL1**	**VGS3**	**VGS4**
Charge	-	0	1	-	0	1
Discharge	-	0	0	-	0	0

Table (2(b)): CRP 1 (Fault at Negative Cycle)

	NORMAL			FAULT		
POSITIVE CYCLE						
MODE	**IL1**	**VGS3**	**VGS4**	**IL1**	**VGS3**	**VGS4**
Charge	+	1	0	+	1	0
Discharge	+	0	0	+	0	0
NEGATIVE CYCLE						
MODE	**IL1**	**VGS3**	**VGS4**	**IL1**	**VGS3**	**VGS4**
Charge	-	0	1	0	0	0
Discharge	-	0	0			

Table 2(c): CRP 2 (Fault at Positive Cycle)

	NORMAL			FAULT		
POSITIVE CYCLE						
MODE	**IL1**	**VGS1**	**VGS2**	**IL1**	**VGS1**	**VGS2**
Charge	+	0	1	0	0	0
Discharge	+	0	0			
NEGATIVE CYCLE						
MODE	**IL1**	**VGS1**	**VGS2**	**IL1**	**VGS1**	**VGS2**
Charge	-	1	0	-	1	0
Discharge	-	0	0	-	0	0

Table 2(d): CRP 2 (Fault at Negative Cycle)

	NORMAL			FAULT		
POSITIVE CYCLE						
MODE	**IL1**	**VGS1**	**VGS2**	**IL1**	**VGS1**	**VGS2**
Charge	+	0	1	+	0	1
Discharge	+	0	0	+	0	0
NEGATIVE CYCLE						
MODE	**IL1**	**VGS1**	**VGS2**	**IL1**	**VGS1**	**VGS2**
Charge	-	1	0	0	0	0
Discharge	-	0	0			

Table 2(a-h) show the values of the gate voltage and the inductor current at positive and negative cycle during charging and discharging process and the identification of fault occurrence for all CRPs. As an example, Table 2(a) shows the fault occurrence at positive cycle for CRP 1. During normal condition at charging state, the gate voltage for S_3 is 15V and gate voltage for S_4 is 0V. This indicates that S_3 is turned-on and S_4 is turned-off at positive cycle. Meanwhile at discharging state, the gate voltage for S_3 and S_4 are 0V, showing that both S_3 and S_4 are turned-off at positive cycle. For inductor current during normal condition at both charging and discharging state, the value of inductor current is positive value. However, during fault condition, the values of inductor current and gate voltages are zero. Compared to negative cycle, the values of inductor current and gate voltages at normal condition and fault condition are similar indicate that the fault occurrence only at positive cycle. The condition is similar for other CRPs.

Table 2(e): CRP 3 (Fault at Positive Cycle)

		NORMAL			FAULT	
	POSITIVE CYCLE					
MODE	IL1	VGS2	VGS4	IL1	VGS2	VGS4
Charge	+	1	0	0	0	0
Discharge	+	0	0			
	NEGATIVE CYCLE					
MODE	IL1	VGS2	VGS4	IL1	VGS2	VGS4
Charge	-	0	1	-	0	1
Discharge	-	0	0	-	0	0

Table 2(f): CRP 2 (Fault at Negative Cycle)

		NORMAL			FAULT	
	POSITIVE CYCLE					
MODE	IL1	VGS2	VGS4	IL1	VGS2	VGS4
Charge	+	1	0	+	1	0
Discharge	+	0	0	+	0	0
	NEGATIVE CYCLE					
MODE	IL1	VGS2	VGS4	IL1	VGS2	VGS4
Charge	-	0	1	0	0	0
Discharge	-	0	0			

Table 2(g): CRP 4 (Fault At Positive Cycle)

		NORMAL			FAULT	
	POSITIVE CYCLE					
MODE	IL1	VGS1	VGS3	IL1	VGS1	VGS3
Charge	+	0	1	0	0	0
Discharge	+	0	0			
	NEGATIVE CYCLE					
MODE	IL1	VGS1	VGS3	IL1	VGS1	VGS3
Charge	-	1	0	-	1	0
Discharge	-	0	0	-	0	0

Table 2(h): CRP 4 (Fault At Negative Cycle)

		NORMAL			FAULT	
	POSITIVE CYCLE					
MODE	IL1	VGS1	VGS3	IL1	VGS1	VGS3
Charge	+	0	1	+	0	1
Discharge	+	0	0	+	0	0
	NEGATIVE CYCLE					
MODE	IL1	VGS1	VGS3	IL1	VGS1	VGS3
Charge	-	1	0	0	0	0
Discharge	-	0	0			

Acknowledgement

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5. Conclusions

A new open-circuit fault detection technique for FTBBR has been proposed in this paper. The detection technique can identify the exact location of the faulty switches under open-circuit fault occurrence by analyzing the output voltage, cycle of the input supply and inductor current. The detection method is simple enough to be practically implemented.